



GY 3µA I_Q, 20mA Low Dropout Linear Regulators

FEATURES

- Ultralow Quiescent Current: 3µA
- Input Voltage Range: 1.6V to 20V
- Output Current: 20mA
- Dropout Voltage: 280mV
- Adjustable Output (V_{ADJ} = V_{OUT(MIN)} = 600mV)
- Fixed Output Voltages: 1.8V, 3.3V, 5V
- Output Tolerance: ±2% Over Load, Line and Temperature
- Stable with Low ESR, Ceramic Output Capacitors (1µF minimum)
- Shutdown Current: <1µA</p>
- Current Limit Protection
- Reverse-Battery Protection
- Thermal Limit Protection
- 8-Lead SC70 and 2mm × 2mm DFN Packages

APPLICATIONS

- Low Current Battery-Powered Systems
- Keep-Alive Power Supplies
- Remote Monitoring Utility Meters Hotel Door Locks

DESCRIPTION

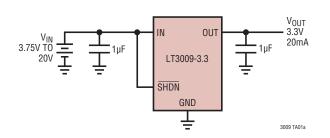
The LT®3009 Series are micropower, low dropout voltage (LDO) linear regulators. The devices supply 20mA output current with a dropout voltage of 280mV. No-load quiescent current is 3μ A. Ground pin current remains at less than 5% of output current as load increases. In shutdown, quiescent current is less than 1 μ A.

The LT3009 regulators optimize stability and transient response with low ESR ceramic capacitors, requiring a minimum of only 1 μ F. The regulators do not require the addition of ESR as is common with other regulators. Internal protection circuitry includes current limiting, thermal limiting, reverse-battery protection and reverse-current protection.

The LT3009 Series are ideal for applications that require moderate output drive capability coupled with ultralow standby power consumption. The device is available in fixed output voltages of 1.8V, 3.3V and 5V, and as an adjustable device with an output voltage range down to the 600mV reference. The LT3009 is available in the 6-lead DFN and 8-lead SC70 packages.

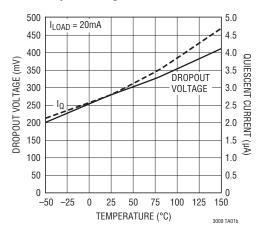
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TYPICAL APPLICATION



3.3V, 20mA Supply with Shutdown

Dropout Voltage/Quiescent Current



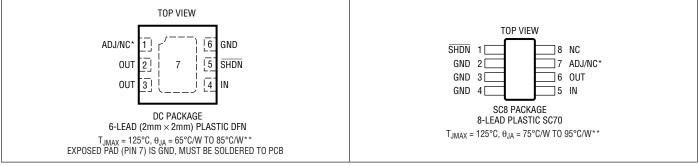
ABSOLUTE MAXIMUM RATINGS

(Note 1)

IN Pin Voltage	±22V
OUT Pin Voltage	±22V
Input-to-Output Differential Voltage	
ADJ Pin Voltage	±22V
SHDN Pin Voltage (Note 8)	±22V
Output Short-Circuit Duration	

Operating Junction Temperature Range (Notes 2, 3)
LT3009E–40°C to 125°C
Storage Temperature Range65°C to 150°C
Lead Temperature: Soldering, 10 sec
SC8 Package Only

PIN CONFIGURATION



* The ADJ pin is not connected in fixed output voltage versions.

** See the Applications Information section.

ORDER INFORMATION

LEAD FREE FINISH	TAPE AND REEL	PART MARKING	PACKAGE DESCRIPTION	TEMPERATURE RANGE
LT3009EDC#PBF	LT3009EDC#TRPBF	LCQX	6-Lead (2mm × 2mm) Plastic DFN	-40°C to 125°C
LT3009EDC-1.8#PBF	LT3009EDC-1.8#TRPBF	LDKC	6-Lead (2mm × 2mm) Plastic DFN	-40°C to 125°C
LT3009EDC-3.3#PBF	LT3009EDC-3.3#TRPBF	LDKD	6-Lead (2mm × 2mm) Plastic DFN	-40°C to 125°C
LT3009EDC-5#PBF	LT3009EDC-5#TRPBF	LDKF	6-Lead (2mm × 2mm) Plastic DFN	-40°C to 125°C
LT3009ESC8#PBF	LT3009ESC8#TRPBF	LCQY	8-Lead Plastic SC70	-40°C to 125°C
LT3009ESC8-1.8#PBF	LT3009ESC8-1.8#TRPBF	LDKG	8-Lead Plastic SC70	-40°C to 125°C
LT3009ESC8-3.3#PBF	LT3009ESC8-3.3#TRPBF	LDKH	8-Lead Plastic SC70	-40°C to 125°C
LT3009ESC8-5#PBF	LT3009ESC8-5#TRPBF	LDKJ	8-Lead Plastic SC70	-40°C to 125°C
LEAD BASED FINISH	TAPE AND REEL	PART MARKING	PACKAGE DESCRIPTION	TEMPERATURE RANGE
LT3009EDC	LT3009EDC#TR	LCQX	6-Lead (2mm × 2mm) Plastic DFN	-40°C to 125°C
LT3009EDC-1.8	LT3009EDC-1.8#TR	LDKC	6-Lead (2mm × 2mm) Plastic DFN	-40°C to 125°C
LT3009EDC-3.3	LT3009EDC-3.3#TR	LDKD	6-Lead (2mm × 2mm) Plastic DFN	-40°C to 125°C
LT3009EDC-5	LT3009EDC-5#TR	LDKF	6-Lead (2mm × 2mm) Plastic DFN	-40°C to 125°C
LT3009ESC8	LT3009ESC8#TR	LCQY	8-Lead Plastic SC70	-40°C to 125°C
LT3009ESC8-1.8	LT3009ESC8-1.8#TR	LDKG	8-Lead Plastic SC70	-40°C to 125°C





ORDER INFORMATION

LEAD FREE FINISH	TAPE AND REEL	PART MARKING	PACKAGE DESCRIPTION	TEMPERATURE RANGE
LT3009ESC8-3.3	LT3009ESC8-3.3#TR	LDKH	8-Lead Plastic SC70	-40°C to 125°C
LT3009ESC8-5	LT3009ESC8-5#TR	LDKJ	8-Lead Plastic SC70	-40°C to 125°C

Consult LTC Marketing for parts specified with wider operating temperature ranges.

For more information on lead free part marking, go to: http://www.linear.com/leadfree/

For more information on tape and reel specifications, go to: http://www.linear.com/tapeandreel/

ELECTRICAL CHARACTERISTICS The \bullet denotes the specifications which apply over the full operating temperature range, otherwise specifications are at $T_J = 25^{\circ}C$. (Note 2)

PARAMETER	CONDITIONS		MIN	ТҮР	MAX	UNITS
Operating Voltage			1.6		20	V
Regulated Output Voltage (Note 4)	LT3009-1.8: V _{IN} = 2.3V, I _{LOAD} = 100µA 2.3V < V _{IN} < 20V, 1µA < I _{LOAD} < 20mA	•	1.782 1.764	1.8 1.8	1.818 1.836	V V
	LT3009-3.3: $V_{IN} = 3.8V$, $I_{LOAD} = 100\mu A$ 3.8V < $V_{IN} < 20V$, 1 $\mu A < I_{LOAD} < 20MA$	•	3.267 3.234	3.3 3.3	3.333 3.366	V V
	LT3009-5: V _{IN} = 5.5V, I _{LOAD} = 100 μ A 3.8V < V _{IN} < 20V, 1 μ A < I _{LOAD} < 20mA	•	4.950 4.900	5 5	5.050 5.100	V V
ADJ Pin Voltage (Notes 3, 4)	V _{IN} = 1.6V, I _{LOAD} = 100µA 1.6V < V _{IN} < 20V, 1µA < I _{LOAD} < 20mA	•	594 588	600 600	606 612	mV mV
Line Regulation (Note 3)	$ \begin{array}{c} \text{LT3009-1.8:} \Delta V_{\text{IN}} = 2.3 \text{V to } 20 \text{V}, \ \text{I}_{\text{LOAD}} = 1 \text{mA} \\ \text{LT3009-3.3:} \Delta V_{\text{IN}} = 3.8 \text{V to } 20 \text{V}, \ \text{I}_{\text{LOAD}} = 1 \text{mA} \\ \text{LT3009-5:} \Delta V_{\text{IN}} = 5.5 \text{V to } 20 \text{V}, \ \text{I}_{\text{LOAD}} = 1 \text{mA} \\ \text{LT3009:} \Delta V_{\text{IN}} = 1.6 \text{V to } 20 \text{V}, \ \text{I}_{\text{LOAD}} = 1 \text{mA} \end{array} $	• • •		1.2 2.2 3.33 0.4	4.5 8.25 12.5 1.5	mV mV mV mV
Load Regulation (Note 3)	$\begin{array}{c} \mbox{LT3009-1.8:} \ \ \ \ \ \ \ \ \ \ \ \ \ \ \ \ \ \ \$	LT3009-1.8: $V_{IN} = 2.3V$, $I_{LOAD} = 1\mu A$ to 20mA LT3009-3.3: $V_{IN} = 3.8V$, $I_{LOAD} = 1\mu A$ to 20mA LT3009-5: $V_{IN} = 5.5V$, $I_{LOAD} = 1\mu A$ to 20mA		2.10 3.85 5.83 0.7	9.00 16.50 25.00 3	mV mV mV mV
Dropout Voltage V _{IN} = V _{OUT(NOMINAL)} (Notes 5, 6)	$I_{LOAD} = 100\mu A$ $I_{LOAD} = 100\mu A$	•		115	180 250	mV mV
	I _{LOAD} = 1mA I _{LOAD} = 1mA	•		170	250 350	mV mV
	$I_{LOAD} = 10mA$ $I_{LOAD} = 10mA$	•		250	310 410	mV mV
	$I_{LOAD} = 20mA$ $I_{LOAD} = 20mA$	•		280	350 450	mV mV
Quiescent Current (Notes 6, 7)	$I_{LOAD} = 0\mu A$ $I_{LOAD} = 0\mu A$	•		3	6	μA μA
GND Pin Current V _{IN} = V _{OUT(NOMINAL)} + 0.5V (Notes 6, 7)	$ \begin{array}{c} I_{LOAD} = 0 \mu A \\ I_{LOAD} = 100 \mu A \\ I_{LOAD} = 1 m A \\ I_{LOAD} = 10 m A \\ I_{LOAD} = 20 m A \end{array} $			3 6 23 250 450	6 12 50 500 1000	μΑ μΑ μΑ μΑ
Output Voltage Noise (Note 9)	$C_{OUT} = 1\mu$ F, $I_{LOAD} = 20$ mA, BW = 10Hz to 100kHz			150		μV _{RMS}
ADJ Pin Bias Current	•		-10	0.3	10	nA
Shutdown Threshold	$V_{OUT} = Off to On$ $V_{OUT} = On to Off$		0.2	0.66 0.36	1.5	V V
SHDN Pin Current	$V_{\overline{SHDN}} = 0V, V_{IN} = 20V$ $V_{\overline{SHDN}} = 20V, V_{IN} = 20V$			500	±1 1600	μA nA 3009fa



ELECTRICAL CHARACTERISTICS The \bullet denotes the specifications which apply over the full operating temperature range, otherwise specifications are at $T_{i} = 25^{\circ}$ C (Note 2)

temperature range, otherwise specifications are at $T_J = 25^{\circ}C$. (Note 2)

PARAMETER CONDITIONS		MIN	ТҮР	MAX	UNITS	
Quiescent Current in Shutdown	$V_{IN} = 6V, V_{\overline{SHDN}} = 0V$	•			<1	μA
Ripple Rejection (Note 3)	$V_{IN} - V_{OUT} = 1.5V$, $V_{RIPPLE} = 0.5V_{P-P}$, $f_{RIPPLE} = 120Hz$, $I_{LOAD} = 20mA$ LT3009 LT3009-1.8 LT3009-3.3 LT3009-5		60 54 49 44	72 66 61 56		dB dB dB dB
Current Limit	$V_{IN} = 20V$, $V_{OUT} = 0$ $V_{IN} = V_{OUT(NOMINAL)} + 1V$, $\Delta V_{OUT} = -5\%$	•	22	60		mA mA
Input Reverse Leakage Current	$V_{IN} = -20V, V_{OUT} = 0$	•		200	350	μA
Reverse Output Current	$V_{OUT} = 1.2V, V_{IN} = 0$			0.6	10	μA

Note 1: Stresses beyond those listed under Absolute Maximum Ratings may cause permanent damage to the device. Exposure to any Absolute Maximum Rating condition for extended periods may affect device reliability and lifetime.

Note 2: The LT3009 is tested and specified under pulse load conditions such that $T_J \cong T_A$. The LT3009 is guaranteed to meet performance specifications from 0°C to 125°C junction temperature. Specifications over the -40° C to 125°C operating junction temperature range are assured by design, characterization and correlation with statistical process controls.

Note 3: The LT3009 adjustable version is tested and specified for these conditions with the ADJ pin connected to the OUT pin.

Note 4: Operating conditions are limited by maximum junction temperature. The regulated output voltage specification will not apply for all possible combinations of input voltage and output current. When operating at the maximum input voltage, the output current range must be limited. When operating at the maximum output current, the input voltage must be limited.

Note 5: Dropout voltage is the minimum input to output voltage differential needed to maintain regulation at a specified output current. In dropout, the output voltage equals ($V_{IN} - V_{DROPOUT}$).

Note 6: To satisfy minimum input voltage requirements, the LT3009 adjustable version is tested and specified for these conditions with an external resistor divider (61.9k bottom, 280k top) which sets V_{OUT} to 3.3V. The external resistor divider adds 9.69µA of DC load on the output. This external current is not factored into GND pin current.

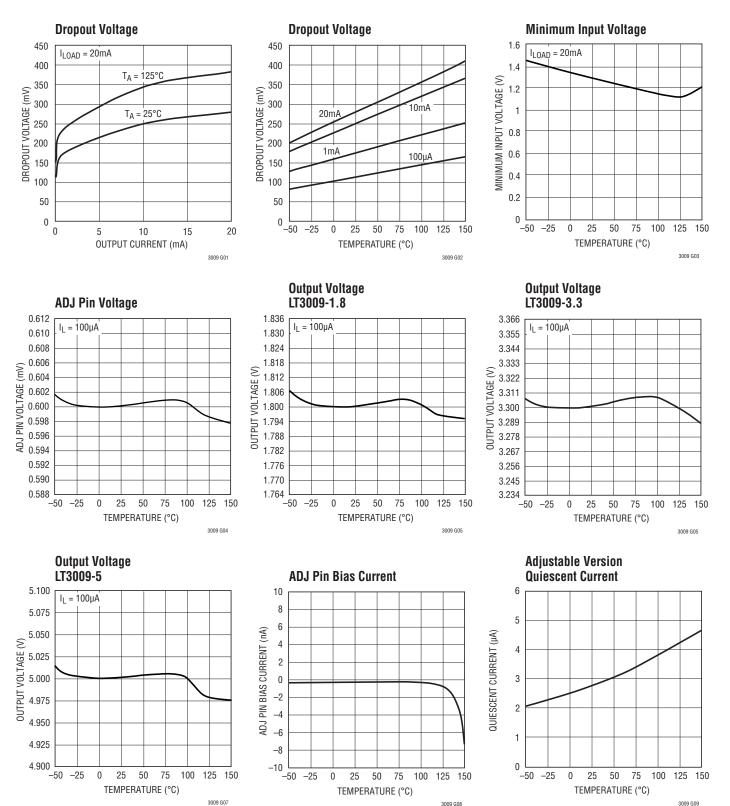
Note 7: GND pin current is tested with $V_{IN} = V_{OUT(NOMINAL)} + 0.5V$ and a current source load. GND pin current will increase in dropout. For the fixed output voltage versions, an internal resistor divider will add to the GND pin current (~2µA for the LT3009-5, ~1µA for the LT3009-1.8 and LT3009-3.3). See the GND Pin Current curves in the Typical Performance Characteristics section.

Note 8: The SHDN pin can be driven below GND only when tied to the IN pin directly or through a pull-up resistor. If the SHDN pin is driven below GND by more than -0.3V while IN is powered, the output will turn on.

Note 9: Output noise is listed for the adjustable version with the ADJ pin connected to the OUT pin. See the RMS Output Noise vs Load Current curve in the Typical Performance Characteristics Section.



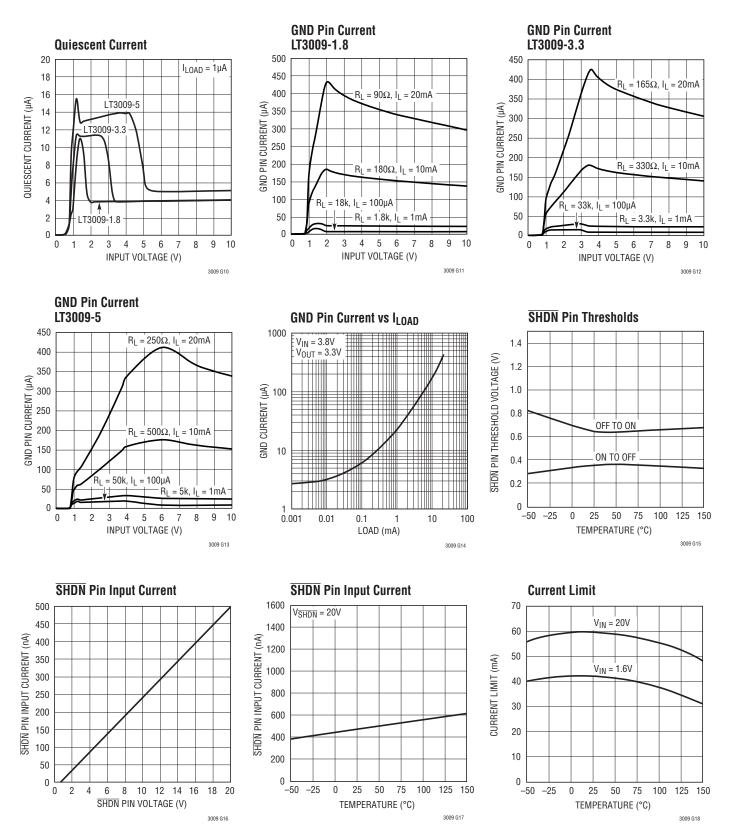
TYPICAL PERFORMANCE CHARACTERISTICS $T_A = 25^{\circ}C$, unless otherwise noted.



9009 GDA

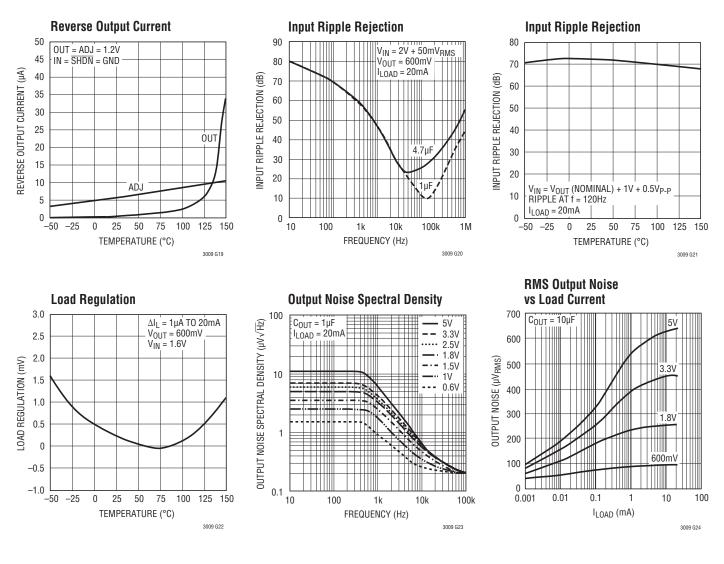


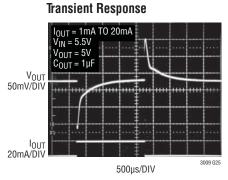
TYPICAL PERFORMANCE CHARACTERISTICS T_A = 25°C, unless otherwise noted.



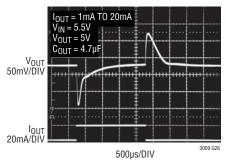


TYPICAL PERFORMANCE CHARACTERISTICS $T_A = 25^{\circ}C$, unless otherwise noted.





Transient Response



PIN FUNCTIONS (SC70/DFN)

SHDN (Pin 1/Pin 5): Shutdown. Pulling the SHDN pin low puts the LT3009 into a low power state and turns the output off. If unused, tie the SHDN pin to V_{IN} . The LT3009 does not function if the SHDN pin is not connected. The SHDN pin cannot be driven below GND unless tied to the IN pin. If the SHDN pin is driven below GND while IN is powered, the output will turn on. SHDN pin logic cannot be referenced to a negative rail.

GND (Pins 2, 3, 4/Pin 6): Ground. Connect the bottom of the resistor divider that sets output voltage directly to GND for the best regulation.

IN (Pin 5/Pin 4): Input. The IN pin supplies power to the device. The LT3009 requires a bypass capacitor at IN if the device is more than six inches away from the main input filter capacitor. In general, the output impedance of a battery rises with frequency, so it is advisable to include a bypass capacitor in battery-powered circuits. A bypass capacitor in the range of 0.1μ F to 10μ F will suffice. The LT3009 withstands reverse voltages on the IN pin with respect to ground and the OUT pin. In the case of a reversed input, which occurs with a battery plugged in backwards, the LT3009 acts as if a large resistor is in series with its input. Limited reverse current flows into the LT3009 and no reverse voltage appears at the load. The device protects both itself and the load.

OUT (Pin 6/Pins 2, 3): Output. This pin supplies power to the load. Use a minimum output capacitor of 1μ F to prevent oscillations. Large load transient applications require larger output capacitors to limit peak voltage transients. See the Applications Information section for more information on output capacitance and reverse output characteristics.

ADJ (Pin 7/Pin 1): Adjust. This pin is the error amplifier's inverting terminal. Its 300pA typical input bias current flows out of the pin (see curve of ADJ Pin Bias Current vs Temperature in the Typical Performance Characteristics section). The ADJ pin voltage is 600mV referenced to GND and the output voltage range is 600mV to 19.5V. This pin is not connected in the fixed output voltage versions.

NC (Pins 7,8/Pin 1): No Connect. For the adjustable voltage version, Pin 8 is an NC pin in the SC70 package. For the fixed voltage versions, Pin 7 and Pin 8 are NC pins in the SC70 package, and Pin 1 is an NC pin in the DFN package. This pin is not tied to any internal circuitry. It may be floated, tied to V_{IN} or tied to GND.

Exposed Pad (Pin 7, DFN Package Only): Ground. The Exposed Pad (backside) of the DFN package is an electrical connection to GND. To ensure optimum performance, solder Pin 7 to the PCB and tie directly to Pin 6.



The LT3009 is a low dropout linear regulator with ultralow quiescent current and shutdown. Quiescent current is extremely low at 3μ A and drops well below 1μ A in shutdown. The device supplies up to 20mA of output current. Dropout voltage at 20mA is typically 280mV. The LT3009 incorporates several protection features, making it ideal for use in battery-powered systems. The device protects itself against both reverse-input and reverse-output voltages. In battery backup applications, where a backup battery holds up the output when the input is pulled to ground, the LT3009 acts as if a blocking diode is in series with its output and prevents reverse current flow. In applications where the regulator load returns to a negative supply, the output can be pulled below ground by as much as 22V without affecting startup or normal operation.

Adjustable Operation

The LT3009 has an output voltage range of 0.6V to 19.5V. Figure 1 shows that output voltage is set by the ratio of two external resistors. The IC regulates the output to maintain the ADJ pin voltage at 600mV referenced to ground. The current in R1 equals 600mV/R1 and the current in R2 is the current in R1 minus the ADJ pin bias current. The ADJ pin bias current, typically 300pA at 25°C, flows out of the pin. Calculate the output voltage using the formula in Figure 1. An R1 value of 619k sets the divider current to 0.97µA. Do not make R1's value any greater than 619k to minimize output voltage errors due to the ADJ pin bias current and to insure stability under minimum load conditions. In shutdown, the output turns off and the divider current is zero. Curves of ADJ Pin Voltage vs Temperature and ADJ Pin Bias Current vs Temperature appear in the Typical Performance Characteristics.

Specifications for output voltages greater than 0.6V are proportional to the ratio of the desired output voltage to 0.6V: $V_{OUT}/0.6V$. For example, load regulation for an output current change of 100µA to 20mA is -0.7mV typical at $V_{OUT} = 0.6V$. At $V_{OUT} = 5V$, load regulation is:

$$\frac{5V}{0.6V} \bullet (-0.7mV) = -5.83mV$$

Table 1 shows resistor divider values for some common output voltages with a resistor divider current of about $1\mu A$.

V _{OUT}	R1	R2		
1V	604k	402k		
1.2V	604k	604k		
1.5V	590k	887k		
1.8V	590k	1.18M		
2.5V	590k	1.87M		
3V	590k	2.37M		
3.3V	619k	2.8M		
5V	590k	4.32M		

Because the ADJ pin is relatively high impedance (depending on the resistor divider used), stray capacitances at this pin should be minimized. Special attention should be given to any stray capacitances that can couple external signals onto the ADJ pin producing undesirable output transients or ripple.

Extra care should be taken in assembly when using high valued resistors. Small amounts of board contamination can lead to significant shifts in output voltage. Appro-

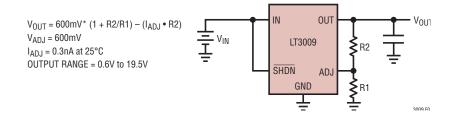


Figure 1. Adjustable Operation

priate post-assembly board cleaning measures should be implemented to prevent board contamination. If the board is to be subjected to humidity cycling or if board cleaning measures cannot be guaranteed, consideration should be given to using resistors an order of magnitude smaller than in Table 1 to prevent contamination from causing unwanted shifts in the output voltage.

Output Capacitance and Transient Response

The LT3009 is stable with a wide range of output capacitors. The ESR of the output capacitor affects stability, most notably with small capacitors. Use a minimum output capacitor of 1µF with an ESR of 3Ω or less to prevent oscillations. The LT3009 is a micropower device and output load transient response is a function of output capacitance. Larger values of output capacitance decrease the peak deviations and provide improved transient response for larger load current changes.

Give extra consideration to the use of ceramic capacitors. Manufacturers make ceramic capacitors with a variety of dielectrics, each with different behavior across temperature and applied voltage. The most common dielectrics are specified with EIA temperature characteristic codes of Z5U, Y5V, X5R and X7R. The Z5U and Y5V dielectrics provide high C-V products in a small package at low cost, but exhibit strong voltage and temperature coefficients as shown in Figures 2 and 3. When used with a 5V regulator, a 16V 10µF Y5V capacitor can exhibit an effective value as low as 1µF to 2µF for the DC bias voltage applied and over the operating temperature range. The X5R and X7R dielectrics yield more stable characteristics and are more suitable for use as the output capacitor. The X7R type has better stability across temperature, while the X5R is less expensive and is available in higher values. One must still exercise care when using X5R and X7R capacitors; the X5R and X7R codes only specify operating temperature range and maximum capacitance change overtemperature. Capacitance change due to DC bias with X5R and X7R capacitors is better than Y5V and Z5U capacitors, but can still be significant enough to drop capacitor values below appropriate levels. Capacitor DC bias characteristics tend to improve as component case size increases, but expected capacitance at operating voltage should be verified.

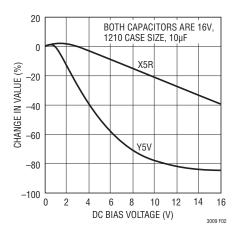


Figure 2. Ceramic Capacitor DC Bias Characteristics

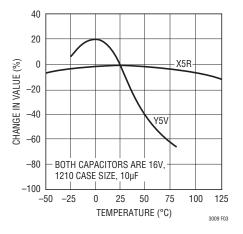


Figure 3. Ceramic Capacitor Temperature Characteristics



Voltage and temperature coefficients are not the only sources of problems. Some ceramic capacitors have a piezoelectric response. A piezoelectric device generates voltage across its terminals due to mechanical stress, similar to the way a piezoelectric accelerometer or microphone works. For a ceramic capacitor, the stress can be induced by vibrations in the system or thermal transients. The resulting voltages produced can cause appreciable amounts of noise, especially when a ceramic capacitor is used for noise bypassing. A ceramic capacitor produced Figure 4's trace in response to light tapping from a pencil. Similar vibration induced behavior can masquerade as increased output voltage noise.

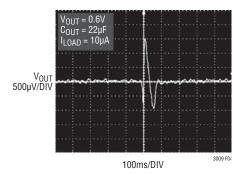


Figure 4. Noise Resulting from Tapping on a Ceramic Capacitor

Thermal Considerations

The LT3009's maximum rated junction temperature of 125°C limits its power-handling capability. Two components comprise the power dissipated by the device:

- 1. Output current multiplied by the input/output voltage differential: $I_{OUT} \bullet (V_{IN} V_{OUT})$
- 2. GND pin current multiplied by the input voltage: $I_{GND} \bullet V_{IN}$

GND pin current is found by examining the GND Pin Current curves in the Typical Performance Characteristics section. Power dissipation is equal to the sum of the two components listed prior.

The LT3009 regulator has internal thermal limiting designed to protect the device during overload conditions. For continuous normal conditions, do not exceed the maximum junction temperature rating of 125°C. Carefully consider all sources of thermal resistance from junction to ambient including other heat sources mounted in proximity to the LT3009. For surface mount devices, heat sinking is accomplished by using the heat spreading capabilities of the PC board and its copper traces. Copper board stiffeners and plated through-holes can also be used to spread the heat generated by power devices.



The following tables list thermal resistance for several different board sizes and copper areas. All measurements were taken in still air on 3/32" FR-4 board with one ounce copper.

COPPER AREA		BOARD	THERMAL RESISTANCE
TOPSIDE*	BACKSIDE	AREA	(JUNCTION-TO-AMBIENT)
2500mm ²	2500mm ²	2500mm ²	65°C/W
1000mm ²	2500mm ²	2500mm ²	70°C/W
225mm ²	2500mm ²	2500mm ²	75°C/W
100mm ²	2500mm ²	2500mm ²	80°C/W
50mm ²	2500mm ²	2500mm ²	85°C/W

Table 2: Measured Thermal Resistance for DC Package

*Device is mounted on the topside.

 Table 3: Measured Thermal Resistance for SC70 Package

COPPE	R AREA	BOARD	THERMAL RESISTANCE
TOPSIDE*	BACKSIDE	AREA	(JUNCTION-TO-AMBIENT)
2500mm ²	2500mm ²	2500mm ²	75°C/W
1000mm ²	2500mm ²	2500mm ²	80°C/W
225mm ²	2500mm ²	2500mm ²	85°C/W
100mm ²	2500mm ²	2500mm ²	90°C/W
50mm ²	2500mm ²	2500mm ²	95°C/W

*Device is mounted on the topside.

Calculating Junction Temperature

Example: Given an output voltage of 3.3V, an input voltage range of $12V \pm 5\%$, an output current range of 0mA to 20mA and a maximum ambient temperature of 85° C, what will the maximum junction temperature be for an application using the DC package?

The power dissipated by the device is equal to:

 $I_{OUT(MAX)} (V_{IN(MAX)} - V_{OUT}) + I_{GND} (V_{IN(MAX)})$

where,

 $I_{OUT(MAX)} = 20mA$

 $V_{IN(MAX)} = 12.6V$

 I_{GND} at (I_{OUT} = 20mA, V_{IN} = 12.6V) = 0.45mA

S0,

P = 20mA(12.6V - 3.3V) + 0.45mA(12.6V) = 191.7mW

The thermal resistance will be in the range of 65°C/W to 85°C/W depending on the copper area. So the junction temperature rise above ambient will be approximately equal to:

0.1917W(75°C/W) = 14.4°C

The maximum junction temperature equals the maximum junction temperature rise above ambient plus the maximum ambient temperature or:

 $T_{J(MAX)} = 85^{\circ}C + 14.4^{\circ}C = 99.4^{\circ}C$



Protection Features

The LT3009 incorporates several protection features that make it ideal for use in battery-powered circuits. In addition to the normal protection features associated with monolithic regulators, such as current limiting and thermal limiting, the device also protects against reverse-input voltages, reverse-output voltages and reverse output-toinput voltages.

Current limit protection and thermal overload protection protect the device against current overload conditions at the output of the device. For normal operation, do not exceed a junction temperature of 125°C.

The LT3009 IN pin withstands reverse voltages of 22V. The device limits current flow to less than 1mA (typically less than 220 μ A) and no negative voltage appears at OUT. The device protects both itself and the load against batteries that are plugged in backwards.

The SHDN pin cannot be driven below GND unless tied to the IN pin. If the SHDN pin is driven below GND while IN is powered, the output will turn on. SHDN pin logic cannot be referenced to a negative rail.

The LT3009 incurs no damage if OUT is pulled below ground. If IN is left open circuit or grounded, OUT can be pulled below ground by 22V. No current flows from the pass transistor connected to OUT. However, current flows

in (but is limited by) the resistor divider that sets output voltage. Current flows from the bottom resistor in the divider and from the ADJ pin's internal clamp through the top resistor in the divider to the external circuitry pulling OUT below ground. If IN is powered by a voltage source, OUT sources current equal to its current limit capability and the LT3009 protects itself by thermal limiting if necessary. In this case, grounding the SHDN pin turns off the LT3009 and stops OUT from sourcing current.

The LT3009 incurs no damage if the ADJ pin is pulled above or below ground by 22V. If IN is left open circuit or grounded, ADJ acts like a 100k resistor in series with a diode when pulled above or below ground.

In circuits where a backup battery is required, several different input/output conditions can occur. The output voltage may be held up while the input is either pulled to ground, pulled to some intermediate voltage or is left open circuit. Current flow back into the output follows the curve shown in Figure 5.

If the LT3009 IN pin is forced below the OUT pin or the OUT pin is pulled above the IN pin, input current typically drops to less than 1μ A. This occurs if the LT3009 input is connected to a discharged (low voltage) battery and either a backup battery or a second regulator circuit holds up the output. The state of the SHDN pin has no effect in the reverse current if OUT is pulled above IN.

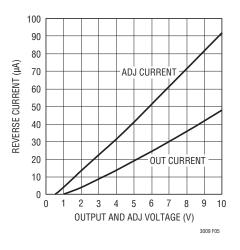
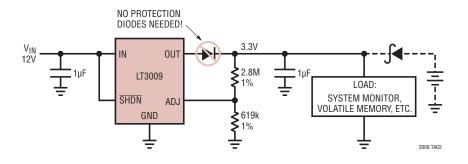


Figure 5. Reverse Output Current

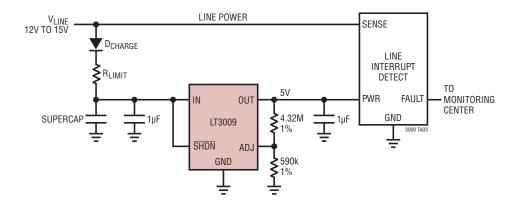


TYPICAL APPLICATIONS

Keep-Alive Power Supply

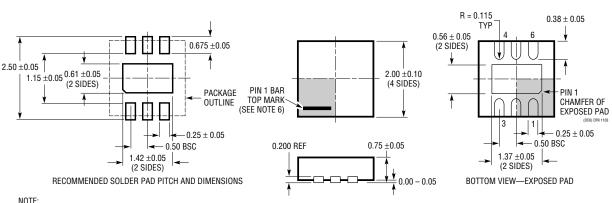


Last-Gasp Circuit





PACKAGE DESCRIPTION



DC Package 6-Lead Plastic DFN (Žmm × 2mm) (Reference LTC DWG # 05-08-1703)

NOTE:

1. DRAWING TO BE MADE A JEDEC PACKAGE OUTLINE M0-229 VARIATION OF (WCCD-2) 2. DRAWING NOT TO SCALE

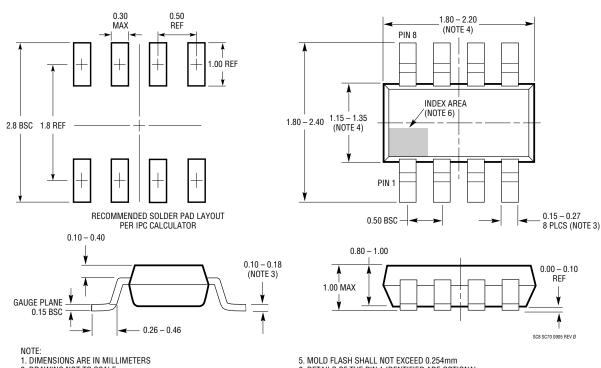
3. ALL DIMENSIONS ARE IN MILLIMETERS

4

DIMENSIONS OF EXPOSED PAD ON BOTTOM OF PACKAGE DO NOT INCLUDE MOLD FLASH. MOLD FLASH, IF PRESENT, SHALL NOT EXCEED 0.15mm ON ANY SIDE

5. EXPOSED PAD SHALL BE SOLDER PLATED 6. SHADED AREA IS ONLY A REFERENCE FOR PIN 1 LOCATION ON THE TOP AND BOTTOM OF PACKAGE

SC8 Package 8-Lead Plastic SC70 (Reference LTC DWG # 05-08-1639 Rev Ø)



- 2. DRAWING NOT TO SCALE
- 4. DIMENSIONS ARE INCLUSIVE OF PLATING 4. DIMENSIONS ARE EXCLUSIVE OF MOLD FLASH AND METAL BURR

6. DETAILS OF THE PIN 1 IDENTIFIER ARE OPTIONAL, BUT MUST BE LOCATED WITHIN THE INDEX AREA

7. EIAJ PACKAGE REFERENCE IS EIAJ SC-70 AND JEDEC MO-203 VARIATION BA



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LT3009 Series

RELATED PARTS

PART NUMBER	DESCRIPTION	COMMENTS		
LT1761	100mA, Low Noise Micropower LDO	V_{IN} : 1.8V to 20V, V_{OUT} = 1.22V, V_{DO} = 0.3V, I_Q = 20µA, I_{SD} < 1µA, Low Noise < 20µV _{RMS} , Stable with 1µF Ceramic Capacitors, ThinSOT TM Package		
LT1762	150mA, Low Noise Micropower LDO	V_{IN} : 1.8V to 20V, V_{OUT} = 1.22V, V_{DO} = 0.3V, I_Q = 25µA, I_{SD} < 1µA, Low Noise < 20µV_{RMS}, MS8 Package		
LT1763	500mA, Low Noise Micropower LDO	V_{IN} : 1.8V to 20V, V_{OUT} = 1.22V, V_{DO} = 0.3V, I_Q = 30µA, I_{SD} < 1µA, Low Noise < 20µV_{RMS}, S8 Package		
LT1764/LT1764A	3A, Low Noise, Fast Transient Response LDOs	V_{IN} : 2.7V to 20V, V_{OUT} = 1.21V, V_{DO} = 0.34V, I_Q = 1mA, I_{SD} < 1µA, Low Noise < 40µV_{RMS}, "A" Version Stable with Ceramic Capacitors, DD and TO220-5 Packages		
LTC1844	150mA, Low Noise Micropower VLDO	V_{IN} : 1.6V to 6.5V, $V_{OUT(MIN)}$ = 1.25V, V_{DO} = 0.09V, I_Q = 35µA, I_{SD} < 1µA, Low Noise: < 30µV_{RMS}, ThinSOT Package		
LT1962	300mA, Low Noise Micropower LDO	V_{IN} : 1.8V to 20V, $V_{OUT(MIN)}$ = 1.22V, V_{D0} = 0.27V, I_Q = 30µA, I_{SD} < 1µA, Low Noise: < 20µV_{RMS}, MS8 Package		
LT1963/LT1963A	1.5A, Low Noise, Fast Transient Response LDOs	V_{IN} : 2.1V to 20V, $V_{OUT(MIN)}$ = 1.21V, V_{D0} = 0.34V, I_Q = 1mA, I_{SD} < 1µA, Low Noise: < 40µV _{RMS} , "A" Version Stable with Ceramic Capacitors, DD, T0220-5, S0T223 and S8 Packages		
LT1964	200mA, Low Noise Micropower, Negative LDO	V_{IN} : –2.2V to –20V, $V_{OUT(MIN)}$ = 1.21V, V_{DO} = 0.34V, I_Q = 30µA, I_{SD} = 3µA, Low Noise: < 30µV _{RMS} , Stable with Ceramic Capacitors,ThinSOT Package		
LT3010	50mA, High Voltage, Micropower LDO	V _{IN} : 3V to 80V, V _{OUT(MIN)} = 1.275V, V _{DO} = 0.3V, I _Q = 30µA, I _{SD} < 1µA, Low Noise: < 100µV _{RMS} , Stable with 1µF Output Capacitor, MS8E Package		
LT3012/LT3012B	250mA, High Voltage, Micropower LDOs	V_{IN} : 4V to 80V, $V_{OUT(MIN)}$ = 1.24V, V_{DO} = 0.4V, I_Q = 40µA, I_{SD} < 1µA, Low Noise: <100µV_{RMS}, Stable with 3.3µF Output Capacitor, 12-Lead 4mm \times 3mm DFN and 16-Lead FE Packages		
LT3013/LT3013B	250mA, High Voltage, Micropower LDOs with PWRGD	V_{IN} : 4V to 80V, $V_{OUT(MIN)}$ = 1.22V, V_{DO} = 0.4V, I_Q = 40µA, I_{SD} < 1µA, Low Noise: < 100µV_{RMS}, Stable with 3.3µF Output Capacitor, 12-Lead 4mm \times 3mm DFN and 16-Lead FE Packages		
LT3014/LT3014B	20mA, High Voltage, Micropower LDO	V_{IN} : 3V to 80V, $V_{OUT(MIN)}$ = 1.2V, V_{DO} = 0.35V, I_Q = 7µA, I_{SD} < 1µA, Low Noise: < 100µV_{RMS}, Stable with 0.47µF Output Capacitor, SOT23-5 and 3mm \times 3mm DFN Packages		
LT3020	100mA, Low Voltage VLDO	V_{IN} : 0.9V to 10V, $V_{OUT(MIN)}$ = 0.20V, V_{DO} = 0.15V, I_Q = 120µA, I_{SD} < 1µA, 3mm \times 3mm DFN and MS8 Packages		
LT3021	500mA, Low Voltage VLDO	V_{IN} : 0.9V to 10V, $V_{OUT(MIN)}$ = 0.20V, V_{DO} = 0.16V, I_Q = 120µA, I_{SD} < 3µA, 5mm \times 5mm DFN and SO8 Packages		
LT3023	Dual 100mA, Low Noise, Micropower LDO	V_{IN} : 1.8V to 20V, $V_{OUT(MIN)}$ = 1.22V, V_{DO} = 0.30V, I_Q = 40µA, I_{SD} < 1µA, DFN and MS10 Packages		
LT3024	Dual 100mA/500mA, Low Noise, Micropower LDO	V_{IN} : 1.8V to 20V, $V_{OUT(MIN)}$ = 1.22V, V_{DO} = 0.30V, I_Q = 60µA, I_{SD} < 1µA, DFN and TSSOP-16E Packages		
LTC3025	300mA, Low Voltage Micropower VLDO	45mV Dropout Voltage, Low Noise 110 μ V _{RMS} , V _{IN} = 1.14V to 5.5V, Low I _Q : 54 μ A, 6-Lead 2mm \times 2mm DFN Package		
LTC3026	1.5A, Low Input Voltage VLDO	100mV Dropout Voltage, Low Noise $80\mu V_{RMS}$, V_{IN} = 0.9V to 5.5V, Low IQ: 950µA, 10-Lead 3mm \times 3mm DFN and MS10E Packages		
LT3027	Dual 100mA, Low Noise, Micropower LDO with Independent Inputs	V_{IN} : 1.8V to 20V, $V_{OUT(MIN)}$ = 1.22V, V_{DO} = 0.30V, I_Q = 40µA, I_{SD} < 1µA, DFN and MS10E Packages		
LT3028	Dual 100mA/500mA, Low Noise, Micropower LDO with Independent Inputs	V_{IN} : 1.8V to 20V, $V_{OUT(MIN)}$ = 1.22V, V_{DO} = 0.30V, I_Q = 60µA, I_{SD} < 1µA, DFN and TSSOP-16E Packages		

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